Error Analysis in an Iterative Algorithm for the Solution of the Regulator Equations for Distributed Parameter Systems

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Statement of Problem

• Consider a MIMO plant modeled by a linear system $z_t(t) = Az(t) + B_{in}u(t) + B_dd(t),$ $z(0) = z_0,$ y(t) = Cz(t).

We are given

- y_r a signal to be tracked,
- *d* a disturbance to be rejected.
- z(t) is the state variable in the infinite dimensional Hilbert space ²
- Problem:

Find the time dependent controller u(t) such that y = Cz(t) satisfies

$$\lim_{t\to\infty}\|y_r(t)-y(t)\|=0.$$

Statement of Problem Cont'd.

- A is an unbounded sectorial operator with dense domain $\mathcal{D}(A)$.
- A generates an exponentially stable analytic semigroup e^{At} in \mathcal{Z}

$$\|e^{At}\| < Me^{-at}$$
 $a > 0$ and $M > 1$.

•
$$B_{\text{in}} \in \mathcal{L}(\mathbb{R}^k, \mathbb{Z})$$
, and $B_{\text{d}} \in \mathcal{L}(\mathbb{R}^m, \mathbb{Z})$.

 C is unbounded but is relatively bounded by a fractional power of (−A). There exists s ∈ (0,2) s.t.

$$C \in \mathcal{L}(\mathcal{H}^{s}, \mathbb{R}^{k}), \text{ where } \mathcal{H}^{s} = \mathcal{D}((-A)^{s/2}).$$

• Without loss of generality, by rescaling, we assume

$$-CA^{-1}B_{\rm in}=I.$$

Dynamic Controller (DC)

• DC

For given $y_r(t)$ and d(t)find \overline{z} , $\overline{\gamma}(t)$ and \overline{z}_0 satisfying $\overline{z}_t = A\overline{z} + B_{in}\overline{\gamma}(t) + B_d d(t)$, $\overline{z}(0) = \overline{z}_0$, such that $C\overline{z}(t) = y_r(t)$, $\forall t \ge 0$.

• In the closed loop system, we set $u(t) = \overline{\gamma}(t)$, and expect $\lim_{t \to \infty} \|Cz(t) - C\overline{z}(t)\| = 0$,

for any initial data z_0 .



Why the Dynamic Controller?

- In the classical geometric method $y_r(t)$ and d(t) are generated by a finite dimensional exogenous system (exosystem)
- In this case we obtain a system of two functional equations called the Regulation Equations
- Solvability of the Regulator Problem is equivalent to the solvability of the Regulator Equations
- Solving the Regulator Equations is not easy.
- The Exosystem might be non-linear.
- The Exosystem might be infinite dimensional.
- The plant could be non-linear.

 $\begin{array}{c} \mbox{Statement of Problem and Dynamic Controller} \\ \beta\mbox{-lterative Scheme} \\ \mbox{Error Estimates} \\ \mbox{Numerical Example} \end{array}$

Regularized Dynamic Controller RDC

• We can rewrite the DC as

$$(I + B_{\text{in}}CA^{-1})\overline{z}_t = A\overline{z}(t) + B_{\text{in}}y_r(t) + (I + B_{\text{in}}CA^{-1})B_{\text{d}}d(t).$$

• For
$$0 we set,$$

 $(I+(1-\beta)B_{\mathrm{in}}CA^{-1})\overline{z}_t^1 = A\overline{z}^1(t) + B_{\mathrm{in}}y_r(t) + (I+B_{\mathrm{in}}CA^{-1})B_{\mathrm{d}}d(t).$

Regularized Dynamic Controller RDC, The Equivalent Form

$$\overline{z}_{t}^{1} = A_{\beta}\overline{z}^{1}(t) + \frac{1}{\beta}B_{in}y_{r}(t) + F d(t),$$

where $A_{\beta} = \left(A - \frac{(1-\beta)}{\beta}B_{in}C\right), \quad F = (I + B_{in}CA^{-1})B_{d}.$
The control can be aposteriori evaluated as

$$\overline{\gamma}^{1}(t) = y_{r}(t) - (1-\beta)CA^{-1}(\overline{z}_{t}^{1}) + CA^{-1}B_{d}d(t).$$

Corollary (Lassi Paunonen)

For $\delta = \frac{1-\beta}{\beta}$ sufficiently close to 0 (thus for β sufficiently close to 1) $A_{\beta} = A - \delta B_{in}C$ generates an exponentially stable semigroup.

General Form of RDC

 We solve the following RDC in each iteration i = 1, 2, ..., for given target y_r(t) and disturbance D(t).

$$\overline{z}_t^i = A_{\beta}\overline{z}^i(t) + \frac{1}{\beta}B_{\rm in}\mathcal{Y}_r(t) + F\mathcal{D}(t).$$

• Then $\overline{\gamma}^{i}(t)$ can be written in following explicit form.

$$\overline{\gamma}^{i}(t) = \mathfrak{Y}_{r}(t) - (1-\beta)CA^{-1}(\overline{z}_{t}^{i}) + CA^{-1}B_{\mathrm{in}}\mathfrak{D}(t).$$

β -iterative scheme

Find approximate values for $\overline{z}(t)$ and $\overline{\gamma}(t)$ by seeking $\overline{z}_n(t) = \sum_{j=1}^n \overline{z}^j(t), \quad \overline{\gamma}_n(t) = \sum_{j=1}^n \overline{\gamma}^j(t)$ with $\overline{z}_n(t) \xrightarrow{n \to \infty} \overline{z}(t)$ and $\overline{\gamma}_n(t) \xrightarrow{n \to \infty} \overline{\gamma}(t)$.

$$E_{1}(t) = y_{r}(t) - C(\overline{z}^{1}(t)),$$

and for $i = 2, 3, ..., n,$
$$E_{i}(t) = E_{i-1}(t) - C(\overline{z}^{i}(t))$$



β -Iterative Algorithm

- Iteration 0: Solve Set Point control problem for z
 ⁰(x) for tracking y_r(0) and rejecting d(0).
- Iteration 1: Solve RDC for $\overline{z}^1(x)$ by setting $\mathcal{Y}_r(t) = y_r(t)$ and $\mathcal{D}(t) = d(t)$ with the I.C. $\overline{z}^1(x, 0) = \overline{z}^0(x)$.
- Iteration i > 1: Solve RDC for $\overline{z}^i(x)$ by setting $\mathcal{Y}_r(t) = E_{i-1}(t)$ and $\mathcal{D}(t) = 0$ with the I.C. $\overline{z}^i(x, 0) = 0$.

Error as a convolution integral.

Let us define

$$\begin{split} \mathcal{K}(t) &= -CA_{\beta}^{-1}e^{A_{\beta}t}\frac{1}{\beta}B_{\text{in}} \text{ and} \\ \mathcal{K}_{d}(t) &= -CA_{\beta}^{-1}e^{A_{\beta}t}(I+B_{\text{in}}CA^{-1})B_{\text{d}}, \end{split}$$

• It can be shown that the first iteration error is given by

$$E_1(t) = y_r(t) - C(\overline{z}^1(t)) = K * y'_r(t) + K_d * d'(t).$$

• and the error at the *i*th-iteration is given by

$$E_i(t) = E_{i-1}(t) - C(\overline{z}^i) = K * E'_{i-1}(t)$$
 for $i = 2, \cdots, n$.

Theorem

Assume $y_r, d \in C_b^n[0,\infty)$ and let β_0 s.t. A_β generates an exponentially stable analytic semigroup in \mathcal{Z} for $\beta \in (\beta_0, 1)$. Then for any T > 0 we have $E_n(t) = \mathcal{E}_{1,T,n}(t) + \mathcal{E}_{2,T,n}(t)$ where $\limsup |\mathcal{E}_{1,T,n}(t)| = 0$ and $t \rightarrow \infty$ $\sup |\mathcal{E}_{2,T,n}(t)|$ $t \in [T,\infty)$ $\leq D^{n}(\sup_{t=1} |y_{r}^{(n)}(t)| + \beta \|B_{in}\|^{-1} \|B_{d}\| \sup_{t=1} |d^{(n)}(t)|),$ $t \in [T,\infty)$ $t \in [T,\infty)$

where $D(A, B_{in}, C, \beta)$ is a constant.

Example 1: 1D Heat Equation with no disturbance

• We consider the control system defined on $0 \le x \le 1$ for $t \ge 0$ given by

$$\frac{\partial z(t)}{\partial t} = Az(t) + B_{in}u(t),$$

$$z(0,t) = 0, \frac{\partial z}{\partial x}(1,t) = 0,$$

$$z(x,0) = 0,$$

$$y(t) = Cz(t), y_r(t) = \cos(t)$$

• In this case we define the operator $A = \frac{d^2}{dx^2}$ with domain $\mathcal{D}(A) = \{\varphi \in H^2(0,1) : \varphi(0) = 0, \quad \varphi'(1) = 0\}$ in the Hilbert state space $\mathcal{Z} = L^2(0,1)$.

•
$$Cz(t) = z(0.75, t)$$
.

•
$$B_{\text{in}}u(t) = \frac{1}{|I_1|}\chi_{I_1}u(t).$$

•
$$I_1 = (0.5 - \delta, 0.5 + \delta).$$



• We set $\delta = 0.05$.

Continuous dependence with respect to β for the eigenvalues of



Figure 1 : Eigenvalues λ_1 , λ_2 , λ_3 , and λ_4 for β varying from 1 to 0.01.

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For $\beta = 0.27$, the plot of $e_1(\text{green}), e_2(\text{blue})$ and $e_3(\text{red})$.



Figure 2 : The first three iteration errors.

$0 \le t \le 20$	Error
$\ e_1\ _{\infty}$	$1.15 imes10^{-1}$
$\ e_2\ _{\infty}$	$4.37 imes10^{-2}$
$\ e_3\ _{\infty}$	$4.36 imes10^{-2}$
$20 \le t \le 35$	Error
$\ e_1\ _{\infty}$	$1.15 imes10^{-1}$
$\ e_2\ _{\infty}$	$1.32 imes 10^{-2}$
$\ e_3\ _{\infty}$	$1.52 imes10^{-3}$
$35 \le t \le 50$	Error
$\ e_1\ _{\infty}$	$1.15 imes10^{-1}$
$\ e_2\ _{\infty}$	$1.32 imes 10^{-2}$
$\ e_3\ _{\infty}$	$1.52 imes 10^{-3}$

 $D\sim 0.1$

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Figure 3 : y_r and C(z).

Example 2: 1D Heat Equation with nonlinear Exosystem

We consider the control system defined on $0 \le x \le 1$ for $t \ge 0$ given by

$$\begin{aligned} \frac{\partial z(t)}{\partial t} &= Az(t) + B_{\text{in}}u(t) + B_{\text{d}}d(t), \\ z(0,t) &= 0, \frac{\partial z}{\partial x}(1,t) = 0, \\ z(x,0) &= 0, \\ y(t) &= Cz(t) = z(1,t). \end{aligned}$$

In this case we define the operator $A = \frac{d^2}{dx^2}$ with domain $\mathcal{D}(A) = \{\varphi \in H^2(0,1) : \varphi(0) = 0, \ \varphi'(1) = 0\}$ in the Hilbert state space $\mathcal{Z} = L^2(0,1)$.

In our specific numerical example we have set

$$I_1 = \{ x : 0 \le x < 1/4 \}, \\ I_2 = \{ x : 1/4 \le x \le 1/2 \},$$



The reference signal y_r and the disturbance d are given by the solution of

$$\ddot{\omega} + \dot{\omega} - \omega + \omega^3 = 0,$$

s.t

for the I.C $\omega(0) = 0, \dot{\omega}(0) = 1.7, \quad y_r(t) = \omega(t) \longrightarrow 1 \text{ as } t \to \infty,$

 $\text{for the I.C } \omega(0)=1, \ \dot{\omega}(0)=1, \quad d(t)=\omega(t), \longrightarrow -1 \text{ as } t \to \infty.$



Figure 4 : The reference y_r and disturbance d.

For $\beta = 0.1$, the plot of $e_1(\text{green}), e_2(\text{blue})$ and $e_3(\text{red})$.



For $\beta = 0.1$



Figure 6 : y_r and C(z)

Example 3: Thermal regulation of a Navier-Stokes Flow

$$\rho(\mathbf{v} \cdot \nabla)\mathbf{v} - \nabla \cdot \left[\mu(\nabla \mathbf{v} + (\nabla \mathbf{v})^{T})\right] + \nabla p = 0, \quad \nabla \cdot \mathbf{v} = 0,$$

$$\frac{\partial T}{\partial t} = \alpha \Delta T - \mathbf{v} \cdot \nabla T + B_{in}u + B_{d}d, \quad y(t) = CT(t),$$

$$y_{r}(t) = a + b\sin(\omega_{1}t), \quad d(t) = c + d\sin(\omega_{2}t)$$

$$\mathbf{v}(x, 0) = 0, p(x, 0) = 0, T(x, 0) = 0.$$

$$\mathbf{v} = 0 \text{ on } \Gamma_{w},$$

$$\mathbf{v} = \begin{pmatrix} f(s) \\ 0 \end{pmatrix}, \quad T = 0 \text{ on } \Gamma_{1},$$

$$-\alpha \nabla T \cdot \mathbf{n} = 0 \text{ on } \Gamma_{2} \cup \Gamma_{w},$$

$$\tau = 0 \text{ on } \Gamma_{2}.$$

Here,

$$\mathcal{A} = \alpha \Delta - \mathbf{v} \cdot \nabla, \ \ \mathcal{D}(\mathcal{A}) = \left\{ \varphi \in \mathcal{H}^2(\Omega) : \varphi \big|_{\Gamma_1} = \mathbf{0}, -\alpha \nabla \varphi \cdot \mathbf{n} \big|_{\Gamma_2 \cup \Gamma_w} = \mathbf{0} \right\}$$

in the Hilbert state space $\mathcal{Z} = L^2(\Omega)$.

$$CT(t) = \frac{1}{|S_3|} \int_{S_3} T ds,$$

$$B_{in}u(t) = rac{1}{|S_1|}\chi_{S_1}u(t), \ \ B_dd(t) = rac{1}{|S_2|}\chi_{S_2}d(t).$$



Figure 7 : The Velocity Profile.

For $\beta = 0.1$



Figure 8 : The first three iteration errors.

 $D \sim 0.3$

$0 \le t \le 100$	Error
$\ e_1\ _{\infty}$	$9.51 imes10^{-1}$
$\ e_2\ _{\infty}$	$5.94 imes10^{-1}$
$\ e_3\ _{\infty}$	$4.72 imes10^{-1}$
$\ e_4\ _{\infty}$	$5.63 imes10^{-1}$
$100 \le t \le 200$	Error
0,	1
€1 ∞	9.49×10^{-1}
$\ e_1\ _{\infty}$	$\begin{array}{c} 9.49 \times 10^{-1} \\ \hline 2.86 \times 10^{-1} \end{array}$
$\ e_1\ _{\infty}$ $\ e_2\ _{\infty}$ $\ e_3\ _{\infty}$	$\begin{array}{c} 9.49 \times 10^{-1} \\ \hline 2.86 \times 10^{-1} \\ \hline 9.86 \times 10^{-2} \end{array}$



Figure 9 : y_r and the measured output of the closed loop system CT.

Future Work

- \bullet Obtain error estimates of the $\beta\mbox{-iterative}$ method for control systems with
 - unbounded B
 - nonlinear state
- Build c++ PDE toolbox for solving Regulator Problem using the β-iterative method

References

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Thank You